

Robust predictive control by zonotopic set-membership estimation

Commande prédictive robuste par des techniques
d'observateurs à base d'ensembles zonotopiques

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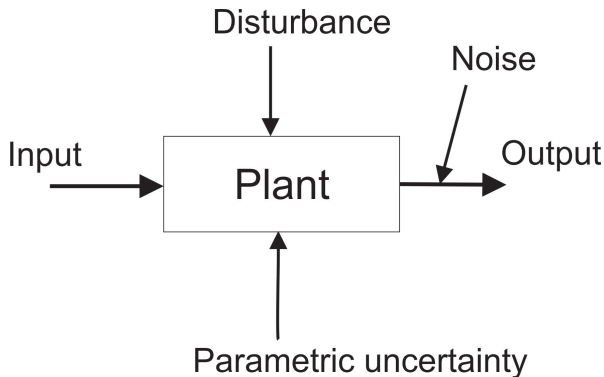
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► Uncertain system



→ Taking into account these uncertainties to the system modeling.

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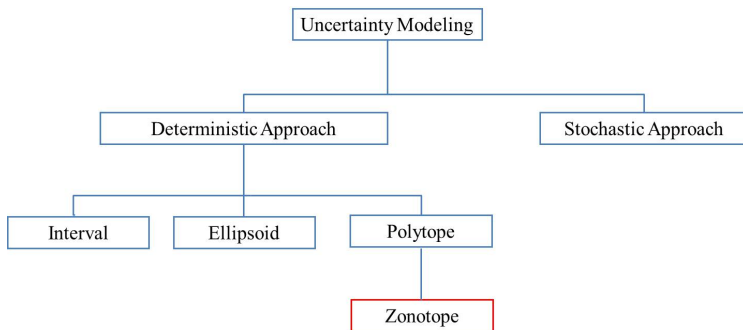
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Uncertainty modeling

- Two approaches for uncertainty modeling



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- ▶ **Interval:** $[a; b] = \{x : a \leq x \leq b\}$
Unitary interval: $\mathbf{B} = [-1; 1]$
Matrix interval: $[A]$ with A_{ij} intervals.
- ▶ **Minkowski sum:** $X \oplus Y = \{x + y : x \in X, y \in Y\}$.
- ▶ **Zonotope:** a convex symmetric polytope
 m -zonotope: the set $p \oplus H\mathbf{B}^m = \{p + Hz : z \in \mathbf{B}^m\}$,
with a vector $p \in \mathbb{R}^n$ and a matrix $H \in \mathbb{R}^{n \times m}$.
- ▶ **P-radius** of a zonotope $Z = p \oplus H\mathbf{B}^m$:
 $L = \max(\|z - p\|_P^2)$, with $z \in Z$ and $P = P^T \succ 0$.

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Example: $p_1 = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$, $H_1 = \begin{bmatrix} 1 & 2 & 3 \\ 3 & 2 & 1 \end{bmatrix}$,

$$p_2 = \begin{bmatrix} 0 \\ 0 \end{bmatrix}, H_2 = \begin{bmatrix} 1 & 0.4 & 3 \\ 3 & 0.2 & 1 \end{bmatrix}, P = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix},$$

$$L_1 = \max_{z \in Z_1} \|z\|_P^2 = 72, L_2 = \max_{z \in Z_2} \|x\|_P^2 = 37.$$

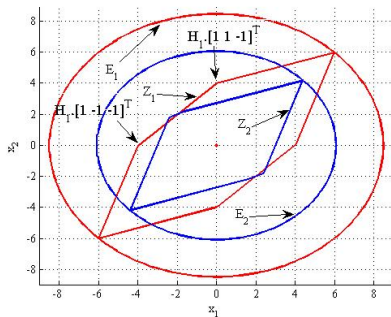


Figure: Zonotopes and ellipsoids representing the associated P -radius

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- ▶ Linear discrete-time system:

$$\begin{cases} x_{k+1} = Ax_k + Bu_k + \omega_k \\ y_k = Cx_k + v_k \end{cases} \quad (1)$$

where

- ▶ $x_k \in \mathbb{R}^{n_x}$ system state vector,
- ▶ $y_k \in \mathbb{R}^{n_y}$ measured output vector,
- ▶ $\omega_k \in \mathbb{R}^{n_x}$ state disturbances,
- ▶ $v_k \in \mathbb{R}^{n_y}$ measurement noise.
- ▶ Assumptions
 1. Detectable, stabilizable.
 2. $\omega_k \in W, v_k \in V$, with W a zonotope, V a box (for simplicity, W, V can be centered in the origin).
 3. x_0 unknown and $x_0 \in X_0$, with X_0 is a zonotope.

Goal: Estimate the system state under uncertainties and **Stabilize** system (1).

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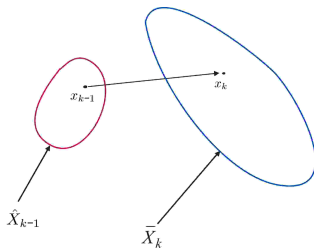
Consider the autonomous system:

$$\begin{cases} x_{k+1} = Ax_k + \omega_k \\ y_k = Cx_k + v_k \end{cases} \quad (2)$$

General algorithm for set-membership estimation:

- Step 1: (*Prediction*)
Compute a set \bar{X}_k that
offers a bound for the
uncertain trajectory of the
system:

$$\bar{X}_k = A\hat{X}_{k-1} \oplus W$$



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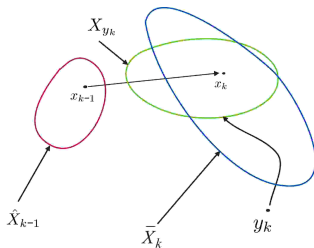
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Consider the autonomous system:

$$\begin{cases} x_{k+1} = Ax_k + \omega_k \\ y_k = Cx_k + v_k \end{cases} \quad (3)$$

General algorithm for set-membership estimation:

- Step 1: (*Prediction*)
- Step 2: (*Measurement*)
Compute the measurement consistent state set X_{y_k} by using the measurement y_k .



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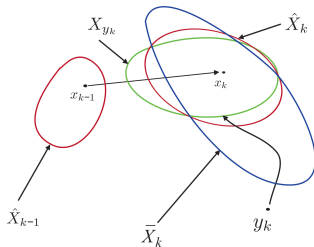
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Consider the autonomous system:

$$\begin{cases} x_{k+1} = Ax_k + \omega_k \\ y_k = Cx_k + v_k \end{cases} \quad (4)$$

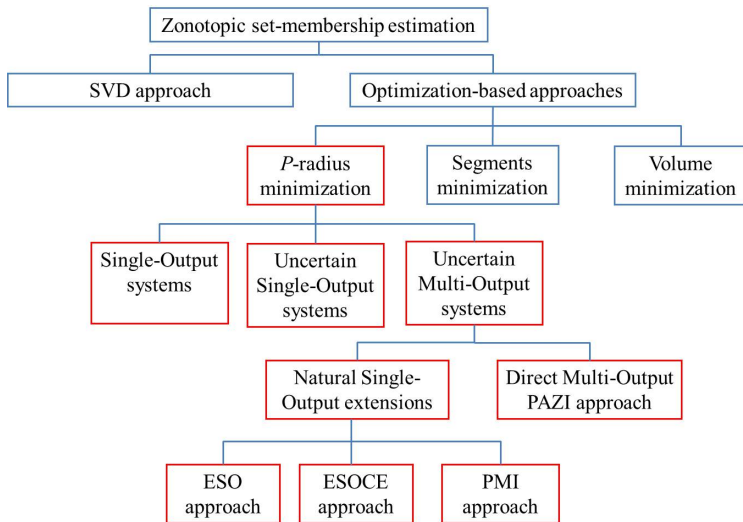
General algorithm for set-membership estimation:

- Step 1: (*Prediction*)
- Step 2: (*Measurement*)
- Step 3: (*Correction*)
Compute an outer approximation \hat{X}_k of the intersection between X_{y_k} and \bar{X}_k .



→ Similar to Kalman filter.

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- Given the system:

$$\begin{cases} x_{k+1} = Ax_k + \omega_k \\ y_k = c^T x_k + v_k \end{cases} \quad (5)$$

with $\omega_k \in W$ a zonotope, $v_k \in V = \sigma \mathbf{B}^1$ an interval.

At the time instant k :

- Guaranteed state set at $k - 1$: $\hat{X}_{k-1} = p \oplus H\mathbf{B}^r$.
- Measured output: y_k .
- Rewrite $W = F\mathbf{B}^{n_x}$.

- Prediction:

$$\bar{X}_k = Ap \oplus [AH \quad F] \mathbf{B}^{r+n_x} \quad (6)$$

- Measurement: $X_{y_k} = \{x \in \mathbb{R}^n : |c^T x - y_k| \leq \sigma\}$
- Correction: State estimation \hat{X}_k the outer approximation of the intersection between a zonotope and a strip.

Intersection of a zonotope and a strip

Property 1: Zonotopic outer approximation of the intersection between a zonotope and a strip¹

Given:

- ▶ zonotope $Z = p \oplus HB^r \subset \mathbb{R}^n$,
- ▶ strip $S = \{x \in \mathbb{R}^n : |c^T x - d| \leq \sigma\}$,
- ▶ vector $\lambda \in \mathbb{R}^n$.

Define:

- ▶ vector $\hat{p}(\lambda) = p + \lambda(d - c^T p) \in \mathbb{R}^n$,
- ▶ matrix $\hat{H}(\lambda) = [(I - \lambda c^T)H \ \sigma \lambda] \in \mathbb{R}^{n \times (m+1)}$.

Then the following expression holds:

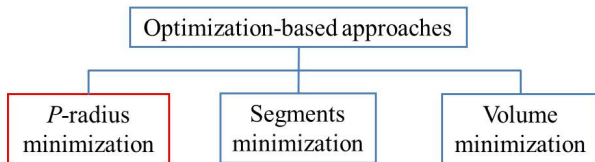
$$Z \cap S \subseteq \hat{Z}(\lambda) = \hat{p}(\lambda) \oplus \hat{H}(\lambda)B^{r+1}.$$

¹Alamo et al. (2005)

Single-Output case

Using Property 1: $\hat{X}_k(\lambda) = \hat{p}(\lambda) \oplus \hat{H}(\lambda) \mathbf{B}^{r+n_x+1}$,
with $\hat{p}(\lambda) = Ap + \lambda(y_k - c^T Ap)$
and $\hat{H}(\lambda) = [(I - \lambda c^T) [AH \ F] \ \sigma\lambda]$

How to choose λ ?



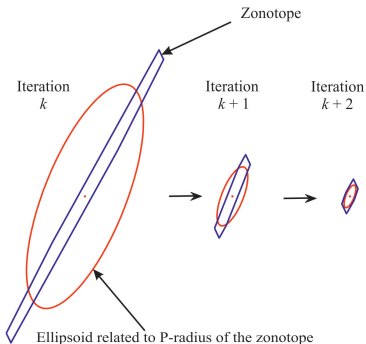
- ▶ Minimizing the segments of the zonotope: simple but not efficient.
- ▶ Minimizing the volume of the zonotope: more accurate and more complex.

→ The proposed P -radius based approach combines the advantages of the two existing approaches.

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New criterion ² to compute λ

Compute a matrix $P = P^T \succ 0$ and a vector λ such that at each sample time, the P -radius of the zonotopic state estimation set is non-increasing.



²Le et al. (2011)

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- The non-increasing condition leads to:

$$\max_{\hat{z} \in \mathbf{B}^{r+n_x+1}} \|\hat{H}\hat{z}\|_P^2 \leq \max_{z \in \mathbf{B}^r} \beta \|Hz\|_P^2 + \max_{s \in \mathbf{B}^{n_x}} \|Fs\|_2^2 + \sigma^2 \quad (7)$$

with $\hat{z} = [z^T \quad s^T \quad \eta]^T$, $\eta \in \mathbf{B}^1$, $\beta \in (0; 1)$.

- Denote by: $L_k = \max_{x \in \hat{X}_k} (\|x - p_k\|_P^2)$ and $\epsilon = \max_{s \in \mathbf{B}^{n_x}} \|Fs\|_2^2$
- Then $(7) \leftrightarrow L_{k+1} \leq \beta L_k + \epsilon + \sigma^2$
- Equivalent BMI (Bilinear Matrix Inequality) problem:

$$\begin{bmatrix} \beta P & 0 & 0 & A^T P - A^T c Y^T \\ * & F^T F & 0 & F^T P - F^T c Y^T \\ * & * & \sigma^2 & Y^T \sigma \\ * & * & * & P \end{bmatrix} \succeq 0 \quad (8)$$

with β , P and $Y = P\lambda$ as decision variables.

Minimization of the P -radius

- ▶ At infinity: $L_\infty = \beta L_\infty + \epsilon + \sigma^2 \Leftrightarrow L_\infty = \frac{\epsilon + \sigma^2}{1 - \beta}$
- ▶ Consider an ellipsoid:

$$E = \{x : x^T P x \leq \frac{\epsilon + \sigma^2}{1 - \beta}\} \Leftrightarrow E = \{x : x^T \frac{(1 - \beta)P}{\epsilon + \sigma^2} x \leq 1\}$$
- ▶ To minimize the size of the guaranteed set, the ellipsoid of the smallest diameter must be found \Rightarrow Eigenvalue Problem (EVP):³

$$\begin{aligned} & \max_{\tau, \beta, P} \tau \\ & \text{subject to BMI} \end{aligned}$$

$$\begin{cases} \tau > 0 \\ P = P^T \succ 0 \\ \frac{(1 - \beta)P}{\epsilon + \sigma^2} \succeq \tau I \end{cases} \quad (9)$$

Diameter of the obtained ellipsoid: $\frac{2}{\sqrt{\tau^*}}$.

³Boyd et al. (1994)

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Global optimization problem

Solve $\max_{\tau, \beta, P, Y} \tau$
subject to BMIs

$$\left\{ \begin{array}{l} \tau > 0 \\ \frac{(1-\beta)P}{\epsilon + \sigma^2} \succeq \tau I \\ \begin{bmatrix} \beta P & 0 & 0 & A^T P - A^T c Y^T \\ * & F^T F & 0 & F^T P - F^T c Y^T \\ * & * & \sigma^2 & Y^T \sigma \\ * & * & * & P \end{bmatrix} \succeq 0 \end{array} \right. \quad (10)$$

-Solution obtained off-line by the Penbmi solver ⁴ or a search loop on $\beta \in (0, 1)$.

-Detectability leads to a feasible solution.

⁴Kočvara and Stingl (2003)

Extension to Single-Output systems with interval uncertainties ⁵

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A is unknown, Schur stable, $A \in [A]$.

Global optimization problem

Solve $\max_{\tau, \beta, P, Y} \tau$
subject to BMIs

$$\left\{ \begin{array}{l} \tau > 0 \\ \frac{(1-\beta)P}{\epsilon + \sigma^2} \succeq \tau I \\ \left[\begin{array}{ccc|cc} \beta P & 0 & 0 & S_i^T P - S_i^T c Y^T & \\ * & F^T F & 0 & F^T P - F^T c Y^T & \\ * & * & \sigma^2 & Y^T \sigma & \\ * & * & * & P & \end{array} \right] \succeq 0 \end{array} \right. \quad (11)$$

with $\beta \in (0, 1)$ and S_i the vertices of $[A]$, $i = 1, \dots, 2^q$ and q the number of interval elements in $[A]$.

⁵Le et al. (2012a)

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Consider the following linear discrete-time invariant system:

$$\begin{cases} x_{k+1} = \begin{bmatrix} 0 & -0.5 \\ 1 & 1 + 0.3\delta \end{bmatrix} x_k + 0.02 \begin{bmatrix} -6 \\ 1 \end{bmatrix} \omega_k \\ y_k = [-2 \quad 1] x_k + 0.2 v_k \end{cases} \quad (12)$$

with $\|v_k\|_\infty \leq 1, \|\omega_k\|_\infty \leq 1$ and $|\delta| \leq 1$.

The initial state belongs to the box $3\mathbf{B}^2$.

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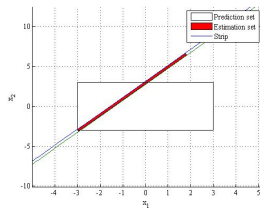
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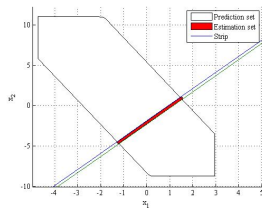
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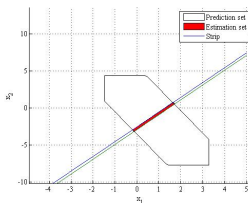
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(a) $k = 1$



(b) $k = 2$



(c) $k = 3$

Figure: Evolution of the guaranteed state estimation \hat{X}_k

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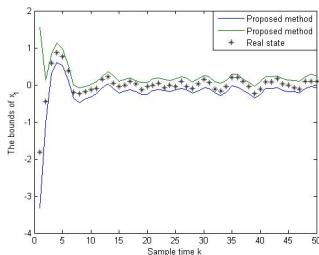
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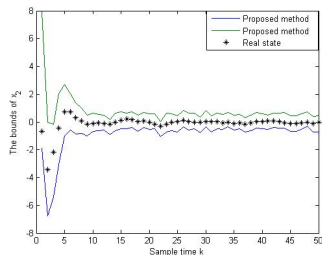
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(a) x_1



(b) x_2

Figure: Guaranteed bounds obtained by proposed method

-The real state is found inside the guaranteed bound
→ good estimation.

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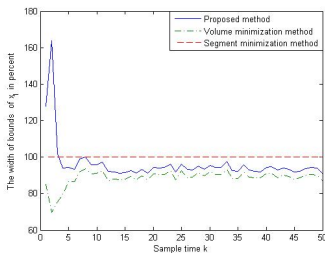
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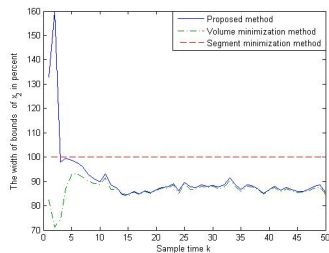
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(a) x_1



(b) x_2

Figure: Comparison of the bound's width obtained by different methods in percent

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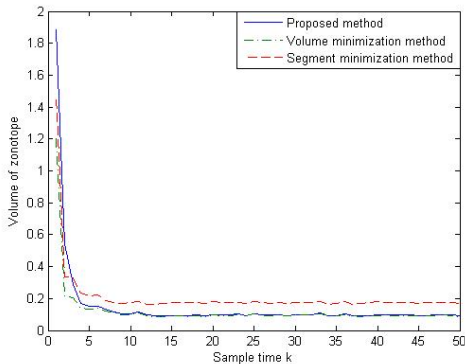


Figure: Comparison of the volume of zonotopic state estimation set obtained by different methods

- The P -radius based approach is better than the segment minimization approach and similar to the volume minimization approach.

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Table: Total computation time after 50 time instants

Approach	Time(second)
Segment minimization	0.0312
P -radius minimization (without off-line optimization (11) included)	0.0312
P -radius minimization (with off-line optimization (11) included)	0.9828
Volume minimization	10.3273

- Conclusion: The P -radius based approach offers a trade-off between the complexity and the precision.

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- Given a system

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with

- $\omega_k \in W$ a zonotope $\subset \mathbb{R}^{n_x}$
- $v_k \in V$ a box $\subset \mathbb{R}^{n_y}$, $V = \Sigma \mathbf{B}^{n_y}$ with $\Sigma = \text{diag}(\sigma_1, \dots, \sigma_{n_y})$

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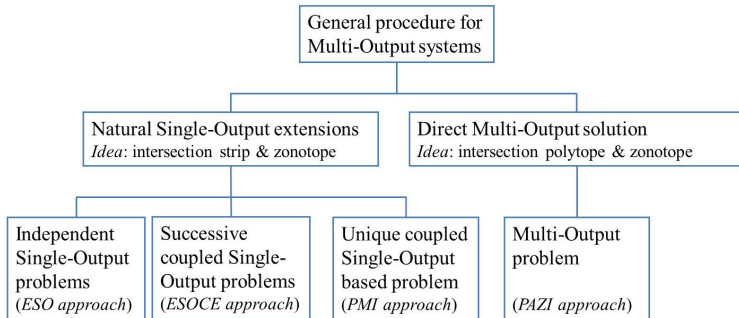
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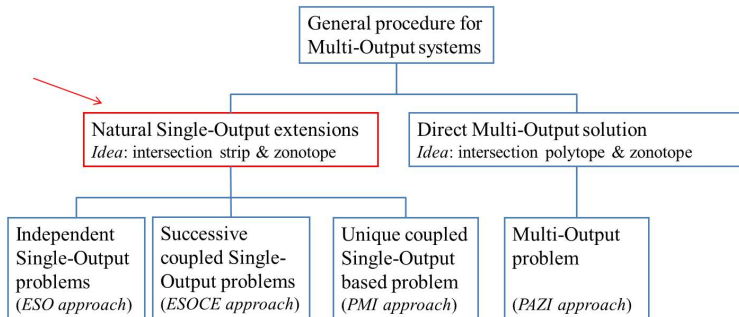
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At the time instant k :

- Guaranteed state set at $k - 1$: $\hat{X}_{k-1} = \hat{p}_{k-1} \oplus \hat{H}_{k-1} \mathbf{B}^r$
- Measured output vector: $y_k = [y_{k/1} \ \dots \ y_{k/n_y}]^T$
- Rewrite $W = F\mathbf{B}^{n_x}$, $V = \text{diag}(\sigma_1, \dots, \sigma_{n_y})\mathbf{B}^{n_y}$

► Prediction:

$$\bar{X}_k = A\hat{p}_{k-1} \oplus [A\hat{H}_{k-1} \ F] \mathbf{B}^{r+n_x} \quad (14)$$

- Measurement: n_y strips $\{x \in \mathbb{R}^n : |c_i^T x - y_{k/i}| \leq \sigma_i\}$,
 $i = 1, \dots, n_y$

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► Correction:

Intersection with the first strip:

$$\hat{X}_{k/1}(\lambda_1) = \hat{p}_{k/1}(\lambda_1) \oplus \hat{H}_{k/1}(\lambda_1) \mathbf{B}^{r+n_x+1} \quad (15)$$

$$\begin{aligned} \text{with } \hat{p}_{k/1}(\lambda_1) &= A\hat{p}_{k-1} + \lambda_1(y_{k/1} - c_1^T A\hat{p}_{k-1}) \\ \hat{H}_{k/1}(\lambda_1) &= [(I - \lambda_1 c_1^T)A\hat{H}_{k-1} \quad (I - \lambda_1 c_1^T)F \quad \sigma_1 \lambda_1]. \end{aligned}$$

Intersection with the second strip:

$$\hat{X}_{k/2}(\lambda_1, \lambda_2) = \hat{p}_{k/2}(\lambda_1, \lambda_2) \oplus \hat{H}_{k/2}(\lambda_1, \lambda_2) \mathbf{B}^{r+n_x+2} \quad (16)$$

$$\begin{aligned} \text{with } \hat{p}_{k/2}(\lambda_1, \lambda_2) &= \hat{p}_{k/1}(\lambda_1) + \lambda_2(y_{k/2} - c_2^T \hat{p}_{k/1}(\lambda_1)) \text{ and} \\ \hat{H}_{k/2}(\lambda_1, \lambda_2) &= [(I - \lambda_2 c_2^T)\hat{H}_{k/1}(\lambda_1) \quad \sigma_2 \lambda_2]. \end{aligned}$$

Correction step

\vdots

until the n_y^{th} strip:

$$\begin{aligned}\hat{X}_{k/n_y}(\lambda_1, \dots, \lambda_{n_y}) &= \hat{p}_{k/n_y}(\lambda_1, \dots, \lambda_{n_y}) \oplus \\ &\oplus \hat{H}_{k/n_y}(\lambda_1, \dots, \lambda_{n_y}) \mathbf{B}^{r+n_x+n_y}\end{aligned}\quad (17)$$

with

$$\begin{aligned}\hat{p}_{k/n_y}(\lambda_1, \dots, \lambda_{n_y}) &= \hat{p}_{k/n_y-1}(\lambda_1, \dots, \lambda_{n_y-1}) + \\ &+ \lambda_{n_y}(y_{k/n_y} - c_{n_y}^T \hat{p}_{k/n_y-1}(\lambda_1, \dots, \lambda_{n_y-1}))\end{aligned}\quad (18)$$

and

$$\begin{aligned}\hat{H}_{k/n_y}(\lambda_1, \dots, \lambda_{n_y}) &= \\ &= \begin{bmatrix} (I - \lambda_{n_y} c_{n_y}^T) \hat{H}_{k/n_y-1}(\lambda_1, \dots, \lambda_{n_y-1}) & \sigma_{n_y} \lambda_{n_y} \end{bmatrix}\end{aligned}\quad (19)$$

→ Guaranteed state estimation at k : $\hat{X}_k = \hat{p}_k \oplus \hat{H}_k \mathbf{B}^{r+n_x+n_y}$
with $\hat{p}_k = \hat{p}_{k/n_y}$, $\hat{H}_k = \hat{H}_{k/n_y}$.

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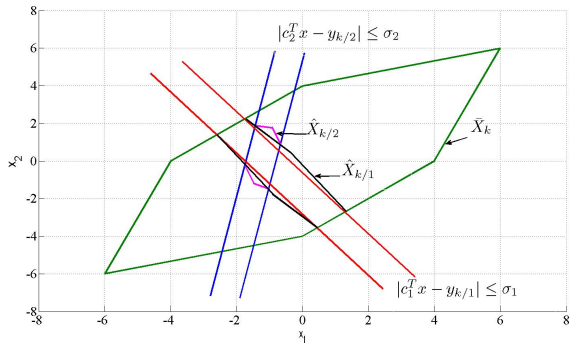
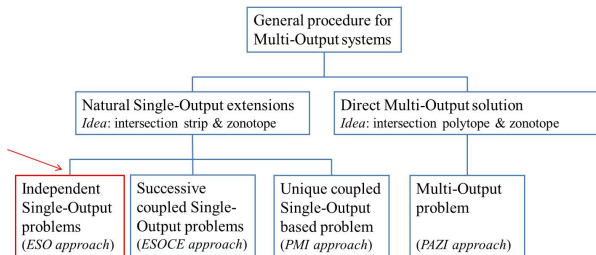


Figure: State estimation of the 2-output system

Equivalent Single-Output approach



Approach 1 to compute λ_i : ESO approach

1. For $j = 1, \dots, n_y$

Step j : Using the strip of the measurement $y_{k/j}$ compute λ_j by solving (10);
End.

2. The guaranteed state estimation is computed by the equation (17) with the known vectors $\lambda_1, \dots, \lambda_{n_y}$.

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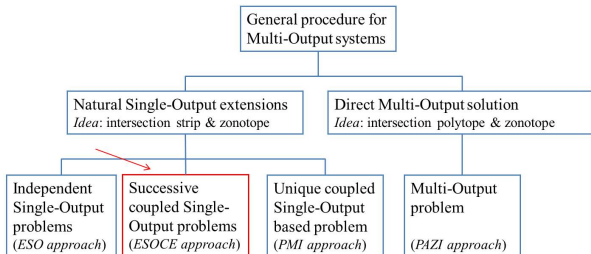
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Equivalent single-output with coupling effect approach



Approach 2 to compute λ_i : ESOCE approach

1. Step 1: Using the measurement $y_{k/1}$ and (10), compute λ_1 ;
2. For $j = 2, \dots, n_y$
Step j : Using the measurement $y_{k/j}$ and the previous obtained vectors $\lambda_1, \dots, \lambda_{j-1}$, compute λ_j by solving (20).
End.

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$$\max_{\tau, \beta, P, Y_j} \tau$$

subject to BMIs

$$\left\{ \begin{array}{l} \tau > 0 \\ \frac{(1-\beta)P}{\sigma_1^2 + \dots + \sigma_j^2 + \epsilon} \succeq \tau I \\ \left[\begin{array}{cccccc} \beta P & 0 & 0 & \dots & 0 & B_1 \\ * & F^T F & 0 & \dots & 0 & B_2 \\ * & * & \sigma_1^2 & \dots & 0 & B_3 \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ * & * & * & \dots & \sigma_j^2 & B_{j+2} \\ * & * & * & \dots & * & P \end{array} \right] \succeq 0 \end{array} \right. \quad (20)$$

with $\beta \in (0, 1)$ and

$$\begin{aligned} B_1 &= ((\prod_{i=1}^j (I - \lambda_{j+1-i} c_{j+1-i}^T)) A)^T P \\ B_2 &= ((\prod_{i=1}^j (I - \lambda_{j+1-i} c_{j+1-i}^T)) F)^T P \\ B_3 &= (\prod_{i=1}^j (I - \lambda_{j+1-i} c_{j+1-i}^T) \sigma_1 \lambda_1)^T P \\ &\vdots \\ B_j &= ((I - \lambda_j c_j^T) (I - \lambda_{j-1} c_{j-1}^T) \sigma_{j-2} \lambda_{j-2})^T P \\ B_{j+1} &= ((I - \lambda_j c_j^T) \sigma_{j-1} \lambda_{j-1})^T P \\ B_{j+2} &= (\sigma_j \lambda_j)^T P \end{aligned} \quad (21)$$

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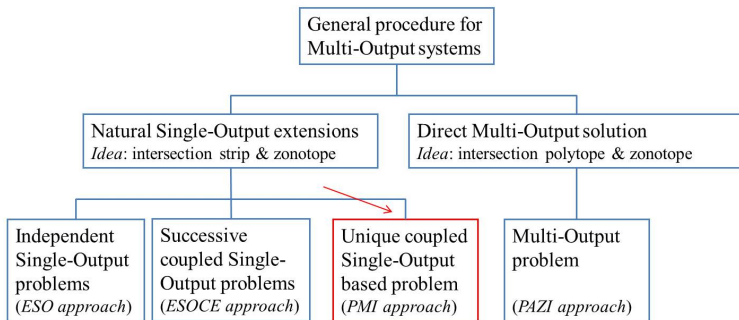
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Polynomial Matrix Inequality approach

Approach 3: PMI approach

Computing all $\lambda_1, \dots, \lambda_{n_y}$ at the same time



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$\max_{\tau, \beta, P, \lambda_1, \dots, \lambda_{n_y}} \tau$
 subject to PMIs

$$\left\{ \begin{array}{l} \tau > 0 \\ \frac{(1-\beta)P}{\sigma_1^2 + \dots + \sigma_{n_y}^2 + \epsilon} \succeq \tau I \\ \left[\begin{array}{cccccc} \beta P & 0 & 0 & \dots & 0 & B_1 \\ * & F^T F & 0 & \dots & 0 & B_2 \\ * & * & \sigma_1^2 & \dots & 0 & B_3 \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ * & * & * & \dots & \sigma_{n_y}^2 & B_{n_y+2} \\ * & * & * & \dots & * & P \end{array} \right] \preceq 0 \end{array} \right. \quad (22)$$

with $\beta \in (0, 1)$ and

$$\begin{aligned} B_1 &= \left(\left(\prod_{i=1}^{n_y} (I - \lambda_{n_y+1-i} c_{n_y+1-i}^T) \right) A \right)^T P \\ B_2 &= \left(\left(\prod_{i=1}^{n_y} (I - \lambda_{n_y+1-i} c_{n_y+1-i}^T) \right) F \right)^T P \\ B_3 &= \left(\prod_{i=1}^{n_y-1} (I - \lambda_{n_y+1-i} c_{n_y+1-i}^T) \sigma_1 \lambda_1 \right)^T P \\ &\vdots \\ B_{n_y} &= \left((I - \lambda_{n_y} c_{n_y}^T) (I - \lambda_{n_y-1} c_{n_y-1}^T) \sigma_{n_y-2} \lambda_{n_y-2} \right)^T P \\ B_{n_y+1} &= \left((I - \lambda_{n_y} c_{n_y}^T) \sigma_{n_y-1} \lambda_{n_y-1} \right)^T P \\ B_{n_y+2} &= (\sigma_{n_y} \lambda_{n_y})^T P \end{aligned} \quad (23)$$

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LMI relaxation solution

PMI problem: difficult to solve → Sub-optimal solution: use of LMI relaxation⁶

Example: $\min_x (-x_1^2 - x_2^2)$

subject to PMI: $\begin{bmatrix} 1 - 4x_1x_2 & x_1 \\ x_1 & 4 - x_1^2 - x_2^2 \end{bmatrix} \succeq 0$

Change of variables:

$y_{10} = x_1$, $y_{01} = x_2$, $y_{20} = x_1^2$, $y_{02} = x_2^2$, $y_{11} = x_1x_2$.

Relaxed optimization problem:

$\min_y (-y_{20} - y_{02})$

subject to LMIs:

$$\left\{ \begin{array}{l} \begin{bmatrix} 1 & * & * \\ y_{10} & y_{20} & * \\ y_{01} & y_{11} & y_{02} \end{bmatrix} \succeq 0 \text{ (Moment matrix)} \\ \begin{bmatrix} 1 - 4y_{11} & * \\ y_{10} & 4 - y_{20} - y_{02} \end{bmatrix} \succeq 0 \end{array} \right. \quad (24)$$

⁶Henrion and Lasserre (2006)

LMI relaxation solution

The same relaxation procedure is applied to the PMI problem (22) with the change of variables:

$$\beta = y_{100\dots 0}, \quad P = \begin{bmatrix} y_{01\dots 0} & \dots & y_{00\dots 1\dots 0} \\ \dots & \dots & \dots \end{bmatrix},$$

$$\lambda_1^T = [y_{00\dots 1\dots 0} \quad \dots], \dots, \lambda_{n_y}^T = [y_{00\dots 1\dots 0} \quad \dots \quad y_{00\dots 01}]$$

Solve the optimization problem

$$\max_{\tau, y} \tau$$

subject to LMIs

$$\left\{ \begin{array}{l} M_1 = \begin{bmatrix} 1 & * & * & * \\ y_{10\dots 0} & y_{20\dots 0} & \dots & * \\ y_{01\dots 0} & y_{11\dots 0} & \dots & * \\ \vdots & \vdots & \ddots & \vdots \\ y_{00\dots 111} & \dots & \dots & y_{00\dots 222} \end{bmatrix} \succeq 0 \\ M_2 \succeq 0 \end{array} \right. \quad (25)$$

M_2 the equivalent LMI expressions obtained from the PMIs in (22) using the new scalar decision variables.

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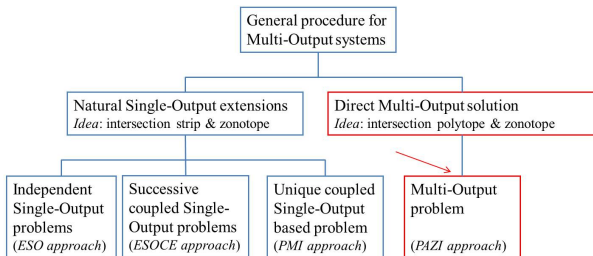
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Direct Multi-Output solution



Previous methods:

- Advantage: direct application of Single-Output case.
- Inconvenient: conservative result due to not computing directly X_{y_k} .

Direct Multi-Output solution:

- Consistent state set created by all the measurements:
$$X_{y_k} = \{x \in \mathbb{R}^{n_x} : |Cx - y_k| \in V\} \Rightarrow \text{a polytope.}$$
- State estimation obtained via the outer approximation of the intersection of zonotope \bar{X}_k and polytope X_{y_k} .

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Proposition: Given:

► zonotope $Z = p \oplus HB^r \subset \mathbb{R}^n$,

► polytope $P = \{x \in \mathbb{R}^n, d \in \mathbb{R}^m : |Cx - d| \leq \begin{bmatrix} \sigma_1 \\ \vdots \\ \sigma_m \end{bmatrix}\}$

$(\sigma_i \in \mathbb{R}^+)$,

► matrix $\Lambda \in \mathbb{R}^{n \times m}$

Define:

► vector $\hat{p}(\Lambda) = p + \Lambda(d - Cp) \in \mathbb{R}^n$,

► matrix $\hat{H}(\Lambda) = [(I - \Lambda C)H \quad \Lambda\Sigma]$, with
 $\Sigma = \text{diag}(\sigma_1, \dots, \sigma_{n_y})$.

Then $Z \cap P \subseteq \hat{Z}(\Lambda) = \hat{p}(\Lambda) \oplus \hat{H}(\Lambda)B^{r+m}$.

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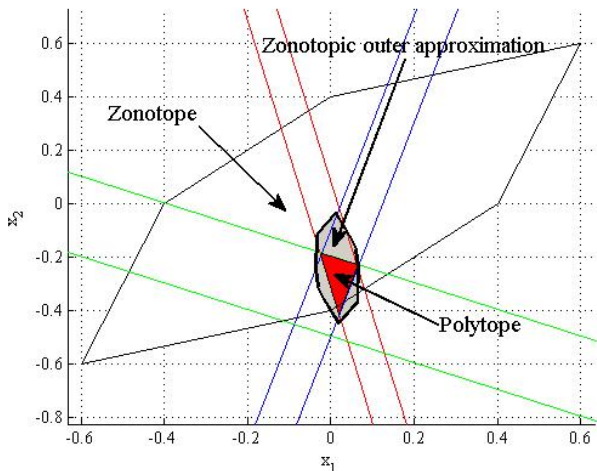


Figure: Zonotopic approximation of the intersection between a zonotope and a polytope

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Similar to Single-Output case:

Based on the estimation at $k - 1$: $\hat{X}_{k-1} = p \oplus H\mathbf{B}^r$,
the state estimation set at k is:

$$\hat{X}_k(\Lambda) = \hat{p}(\Lambda) \oplus \hat{H}(\Lambda)\mathbf{B}^{r+n_x+n_y} \quad (26)$$

with $\hat{p}(\Lambda) = Ap + \Lambda(y_k - CAp)$
and $\hat{H}(\Lambda) = [(I - \Lambda C) \begin{bmatrix} AH & F \end{bmatrix} \quad \Lambda\Sigma]$

► Remark:

- Single-Output case: λ is a vector.
- Multi-Output case: Λ is a matrix.

Direct Multi-Output solution

$\Lambda \in \mathbb{R}^{n_x \times n_y}$ computed to ensure the non-increasing property of the P -radius of the zonotopic guaranteed state estimation:

$\max_{\tau, \beta, P, Y} \tau$
 subject to BMIs

$$\left\{ \begin{array}{l} \tau > 0 \\ \frac{(1-\beta)P}{\sigma_1^2 + \dots + \sigma_{n_y}^2 + \epsilon} \succeq \tau I \\ \begin{bmatrix} \beta P & 0 & 0 & A^T P - A^T C^T Y^T \\ * & F^T F & 0 & F^T P - F^T C^T Y^T \\ * & * & \Sigma^T \Sigma & Y^T \Sigma \\ * & * & * & P \end{bmatrix} \preceq 0 \end{array} \right. \quad (27)$$

with $\beta \in (0, 1)$, P and $Y = P\Lambda$ as decision variables.

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Consider the following linear discrete-time invariant system:

$$\begin{cases} x_{k+1} = \begin{bmatrix} 0 & -0.5 \\ 1 & 1 + 0.3\delta \end{bmatrix} x_k + \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix} \omega_k \\ y_k = \begin{bmatrix} -2 & 1 \\ 1 & 1 \end{bmatrix} x_k + \begin{bmatrix} 0.2 & 0 \\ 0 & 0.2 \end{bmatrix} v_k \end{cases} \quad (28)$$

with $\|v_k\|_\infty \leq 1, \|\omega_k\|_\infty \leq 1, \|\delta\| \leq 1$.

The initial state belongs to the box $3\mathbf{B}^2$.

Simulation results

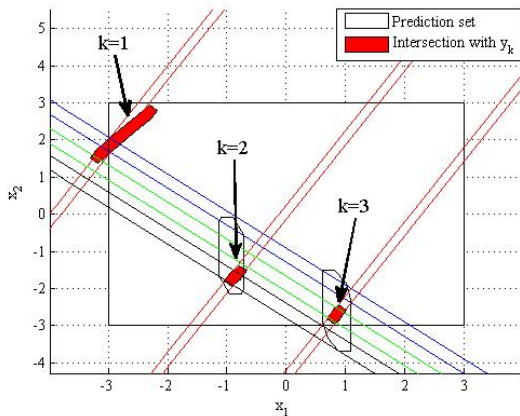


Figure: Evolution of the guaranteed state estimation

-The guaranteed state estimation decreases in time.

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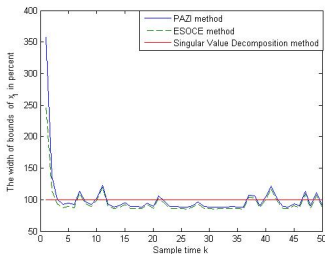
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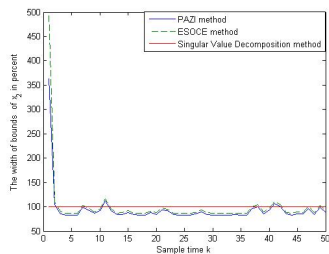
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(a) x_1



(b) x_2

Figure: Comparison of the state bound's width obtained by ESOCE, PAZI and SVD approaches

Simulation results

Table: Total computation time after 50 samples

Approach	Time(second)
PAZI approach (without off-line LMI optimization included)	0.0468
PAZI approach (with off-line LMI optimization included)	0.2808
ESOCE approach (with off-line optimization included)	1.4664
Singular Value Decomposition approach ⁷	1.5444

- Conclusion: PAZI approach offers a low complexity of computation and good precision of estimation.

⁷Combastel (2003)

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Classical output feedback MPC tube-based ⁸:

$$\begin{cases} x_{k+1} = Ax_k + Bu_k + \omega_k \\ y_k = Cx_k + v_k \end{cases} \quad (29)$$

Luenberger observer:

$$\begin{cases} \hat{x}_{k+1} = A\hat{x}_k + Bu_k + L(y_k - \hat{y}_k) \\ \hat{y}_k = C\hat{x}_k \end{cases} \quad (30)$$

Difference equation of state estimation error ($\tilde{x}_k = x_k - \hat{x}_k$):
 $\tilde{x}_{k+1} = (A - LC)\tilde{x}_k + \omega_k^e$, with $A - LC$ stable and
 $\omega_k^e = \omega_k - Lv_k$.

Remark: $\tilde{x}_k \in$ invariant set $S_k^e \rightarrow \tilde{x}_{k+1} \in$ invariant set
 $S_{k+1}^e = (A - LC)S_k^e \oplus W^e$ with $W^e = W \oplus (-LV)$

⁸Mayne et al. (2009)

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Proposed method: Replace Luenberger observer by the zonotopic guaranteed state estimation.

Consider the center of the zonotopic guaranteed state estimation as the estimated state: $\hat{x}_k = p_k$.

Remark: $\tilde{x}_k = x_k - \hat{x}_k \in$ invariant zonotope S_k^e
 $\rightarrow \tilde{x}_{k+1} \in$ invariant $S_{k+1}^e = (I - \Lambda C)AS_k^e \oplus W^e$, with
 $W^e = (I - \Lambda C)W \oplus (-\Lambda V)$

The size of S_k^e decreases by appropriately choosing Λ .

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Nominal system without disturbances:

$$\underline{x}_{k+1} = A\underline{x}_k + B\underline{u}_k \quad (31)$$

To counteract the disturbances, the trajectory is desired to lie close to the nominal trajectory:

$$u_k = \underline{u}_k + Ke_k \quad (32)$$

where $e_k = \hat{x}_k - \underline{x}_k$ satisfies the difference equation:
 $e_{k+1} = (A + BK)e_k + w_k^{co}$, where $A + BK$ stable,
 $w_k^{co} \in W_k^{co} = \Lambda CAS_k^e \oplus \Lambda CW \oplus \Lambda V$

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How to define \underline{u}_k ?

Consider the following cost function at time instant k :

$$V_N(\underline{x}, \underline{u}) = V_f(\underline{x}_N) + \sum_{i=0}^{N-1} l(\underline{x}_i, \underline{u}_i) \quad (33)$$

with $\underline{u} = \{\underline{u}_0, \dots, \underline{u}_{N-1}\}$, $l(\underline{x}, \underline{u}) = 0.5(\underline{x}^T Q \underline{x} + \underline{u}^T R \underline{u})$,
 $V_f(\underline{x}) = 0.5 \underline{x}^T P_f \underline{x}$,

under the following time varying constraints:

$$\begin{cases} \underline{u}_i \in \underline{U}_{k+i}, & i \in \mathbb{N}_{[0, N-1]} \\ \underline{x}_i \in \underline{X}_{k+i}, & i \in \mathbb{N}_{[0, N-1]} \\ \underline{x}_N \in \underline{X}_f \end{cases} \quad (34)$$

Remark: Sets \underline{U}_{k+i} and \underline{X}_{k+i} are time varying due to the decreasing of the estimation set.

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Solution of the optimization problem:

$$(\underline{x}^*(\hat{x}, k), \underline{u}^*(\hat{x}, k)) = \arg \min_{\underline{x}, \underline{u}} \{V_N(\underline{x}, \underline{u})\} \quad (35)$$

The control law applied to the system:

$$\kappa_N(\hat{x}, k) = \hat{u}^*(0, \hat{x}, k) + K(\hat{x}_k - \underline{x}^*(\hat{x}, k)) \quad (36)$$

with $\hat{u}^*(0, \hat{x}, k)$ the first element of the sequence $\underline{u}^*(\hat{x}, k)$.

Using the control law (36), (x, \hat{x}) is robustly steered to some sets containing the origin, exponentially fast satisfying all constraints⁸.

⁸Mayne et al. (2009)

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Consider the following linear discrete-time invariant system:

$$\begin{aligned}x_{k+1} &= \begin{bmatrix} 1 & 1.1 \\ 0 & 1 \end{bmatrix} x_k + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u_k + \omega_k \\ y_k &= \begin{bmatrix} -2 & 1 \end{bmatrix} x_k + v_k\end{aligned}$$

- ▶ $\|v_k\|_\infty \leq 0.05, \|\omega_k\|_\infty \leq 0.1$.
- ▶ $(x, u) \in X \times U$, with:
 $X = \{x \in \mathbb{R}^2 : x_1 \in [-50, 3], x_2 \in [-50, 3]\}$
 $U = \{u \in \mathbb{R} : |u| \leq 9\}$.

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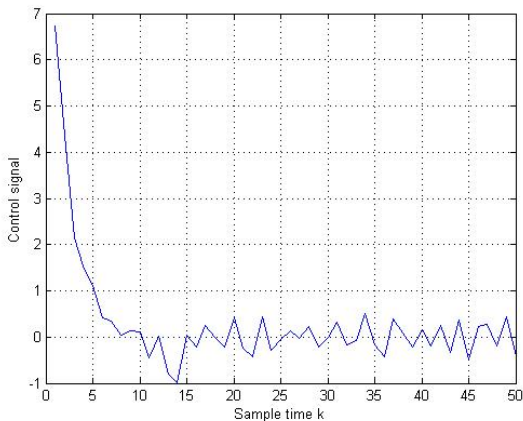


Figure: Control signal

-The control constraint is respected.

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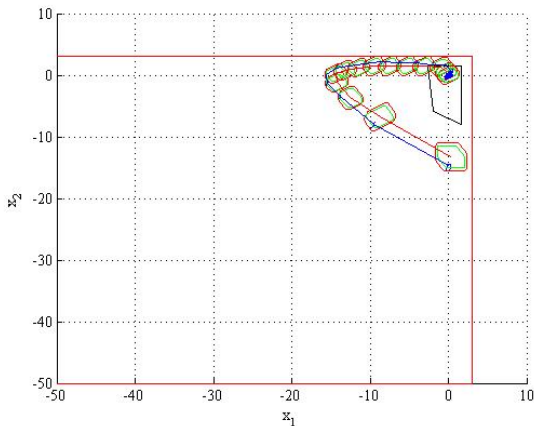


Figure: Tube trajectory of the closed-loop response of the system fulfilling the state constraint

-The state constraints are respected.

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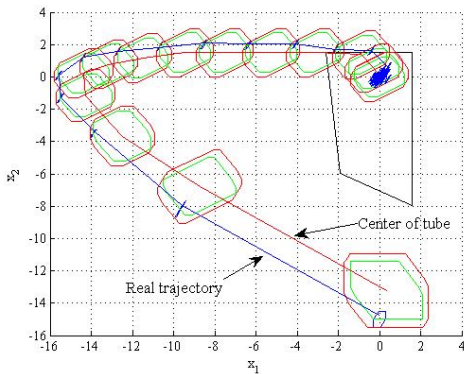


Figure: Zoom of tube trajectory of the closed-loop response of the system

- The size of the tube decreases in time.
- The state converges to an invariant set while respecting the constraints

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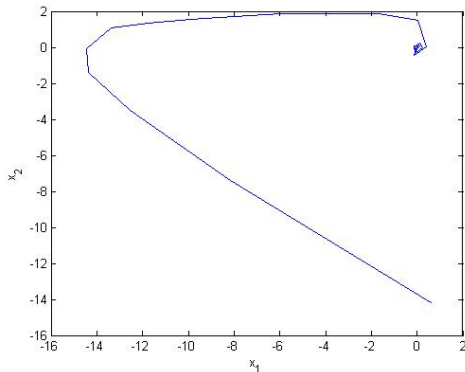


Figure: Closed-loop response of the system

-The system is stable in the sense of Input to State Stability.

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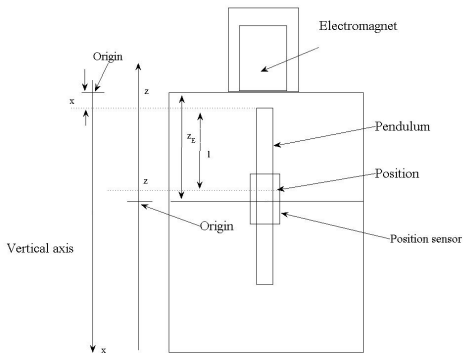


Figure: Magnetic levitation system

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- Consider the continuous-time model⁹:

$$\begin{cases} \dot{x} = \begin{bmatrix} 0 & 1 \\ \frac{2g}{x_0} & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ \frac{g}{i_0} \end{bmatrix} u \\ y = \begin{bmatrix} 1 & 0 \end{bmatrix} x \end{cases} \quad (37)$$

with $g = 9.81m/s^2$, $x_0 = 0.019m$, $i_0 = 0.436A$.

- This model is next discretized with $T_s = 0.1s$.
- Additional model disturbances
 $\omega_k \in W = \{w \in \mathbb{R} : |w| \leq 1\}$ and measurement noise
 $v_k \in V = \{v \in \mathbb{R} : |v| \leq 0.05\}$
- Constraints: $X = \{x \in \mathbb{R}^2 : |x_1| \leq 0.5, |x_2| \leq 10\}$,
 $U = \{u \in \mathbb{R} : |u| \leq 5\}$

⁹Le et al. (2012b)

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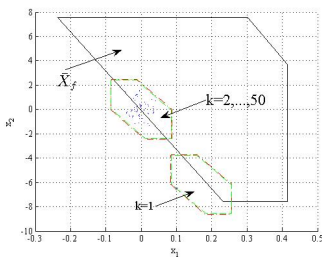
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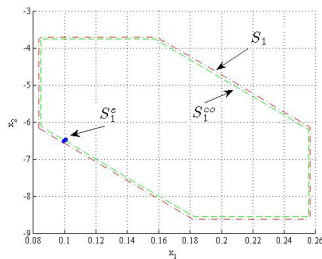
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(a) Tube trajectory



(b) Zoom

Figure: Tube trajectory of the controlled magnetic levitation system

- In this case, the real system enters to the terminal set after two time instants.

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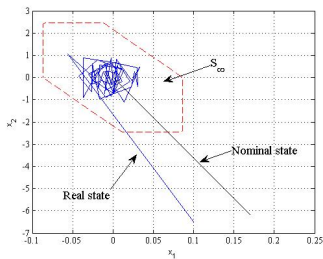
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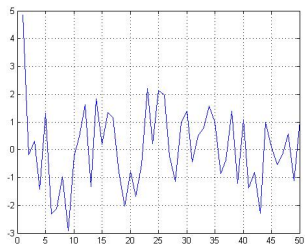
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(a) Real state and nominal state of the closed-loop



(b) Control signal

- Guaranteeing stability of the output feedback system while respecting the constraints.

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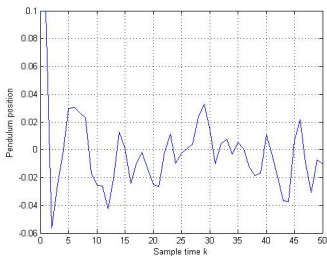
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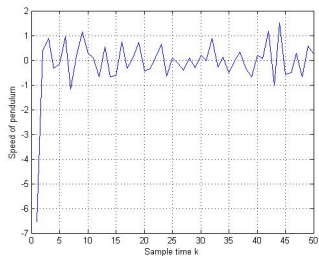
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(c) Pendulum position



(d) Pendulum speed

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► Estimation

1. A new method based on the minimization of the P -radius to solve the problem of set-membership estimation using zonotope is presented.
2. Both Single-Output and Multi-Output systems were considered subject to interval uncertainties, unknown but bounded disturbances and measurement noises.
3. A good trade-off solution between precision and complexity compared to existing zonotopic set-membership estimation techniques is illustrated.

► Control

1. The zonotopic set-membership estimation is used in the context of Tube MPC, offering good performance.

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- ▶ Output feedback control law based on the P -radius based zonotopic set-membership estimation for systems with interval parametric uncertainties in the presence of disturbances, measurement noises and constraints.
- ▶ Extension of the estimation technique to time delay systems.
- ▶ Extension to fault diagnosis problem and fault tolerant control.

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